About the stability of sampled-data systems with non-uniform sampling

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#### Outline

Introduction and Problem Formulation

Existing work

Exponential Stability

Convex embedding

Numerical example

Conclusion

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## Motivating problem : Digital control

Classical control loop



Ideal Hypothesis :

Sampling and actuation are periodic and synchronous

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## Motivating problem : Digital control

Classical control loop



Real-time problem : the system is affected by timing problems

- sampling jitter (sensor, multitasking processors, packet dropouts in communication channnels)
- unknown time varying delays (not adressed here)

(Wittenmark, Nilsson, Torngren, 1995)

#### **Problem Formulation**



Consider the system

$$\dot{x}(t) = Ax(t) + Bu(t), \ \forall t \in \mathbb{R}^+$$

with a sampled-data control :

$$u(t) = Kx(t_k), \ \forall t \in [t_k, t_{k+1})$$

**Problem :** is the system robust to jitter?

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Sampling jitter example (Zhang,2001)

$$\dot{x} = Ax + Bu_k, \ u_k = Kx_k, \ h_k \in \{T_1, T_2\}$$



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Sampling jitter example (Zhang, 2001)  $\Rightarrow$  instability

$$\dot{x} = Ax + Bu_k, \ u_k = Kx_k, \ h_k \in \{T_1, T_2\}$$



**Open problem** : provide tools for robust stability and performance analyzis!

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### Existing work - Continuous-time : time delay approach



 $u(t) = Kx(t_k) = Kx(t - \tau)$  with  $\tau = t - t_k$ ,  $0 < \tau_k < h_{max}$ 

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### Existing work - Continuous-time

- Fridman et al, 2004 (input delay appraoch)
- Mirkin, 2007 (robust control equivalent)
- Hespanha, 2008 (impulsive delay diff. eq.)

#### Advantage :

- Directly extend to performance study (decay rate)
- Take into account the inter-sampling behaviour

#### Inconvenient :

 Do not take into account the sawtooth form of the delay (conservatism)

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#### Existing work - Discrete-time (LPV model)



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## Existing work - Discrete-time (LPV model)

- Sala, 2004; Boyd,2008 (gridding approach)
- Hetel, 2006, 2009 (convex embedding)
- Fujioka, 2007 (gridding + robust control)

Advantage :

 Implicitly take into account the sawtooth form of the delay (less conservative)

#### Inconvenient :

 Not numerically efficient when the minimum sampling interval goes to zero

Ignore the inter-sampling behaviour

## Discrete-time problem (inter-sampling behaviour)



Evolution of a Lyapunov function  $V(x) = x^T P x$ 

- Striclty decreasing at t = t<sub>k</sub> (sufficient condition for stability analyzis)
- Increasing in between the sampling times (false evaluation of control performance)

#### Provide a continuous-time method that takes into account the sawtooth form of the delay (advantage of discrete-time methods for conservatism reduction)

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 $\blacktriangleright$  how fast the norm of the state vector converges to zero.  $\alpha>0$ 

$$\|x(t)\| \leq e^{-\alpha t} c \|x(0)\|, \forall t > 0$$

• for a given candidate Lyapunov function V(x), if

$$\frac{dV(x)}{dt} < -2\alpha V(x), \ \forall x \neq 0$$

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#### Case of quadratic Lyapunov functions

For 
$$\frac{dx(t)}{dt} = Ax(t) + BKx(t-\tau(t)), \ \tau(t) := t-t_k, \ \forall t \in [t_k, t_{k+1})$$

and  $V(x) = x^T P x$ 

Derivative of Lyapunov function

$$\frac{dV(x)}{dt} = 2x^{T}(t)P(Ax(t) + BKx(t-\tau)).$$

Sawtooth evolution of delay can be introduced by using the integration operator Λ(·) used for the discrete-time model :

$$x(t) = \Lambda(t-t_k) x(t_k), \ \forall t \in [t_k, t_{k+1})$$

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$$\mathbf{x}(t) = \Lambda(\tau) \mathbf{x}(t-\tau), \ \forall t \in [t_k, t_{k+1})$$

Estimation of the decay rate

#### $\alpha > 0, P \succ 0$

# $\begin{pmatrix} \mathsf{K}^{\mathsf{T}} \mathsf{B}^{\mathsf{T}} + \mathsf{\Lambda}^{\mathsf{T}}(\tau) \mathsf{A}^{\mathsf{T}} \end{pmatrix} \mathsf{P} \mathsf{\Lambda}(\tau) + \mathsf{\Lambda}^{\mathsf{T}}(\tau) \mathsf{P}(\mathsf{A} \mathsf{\Lambda}(\tau) + \mathsf{B} \mathsf{K}) \\ \prec -2\alpha \mathsf{\Lambda}^{\mathsf{T}}(\tau) \mathsf{P} \mathsf{\Lambda}(\tau) \,,$

 $\forall \tau \in [0, h_{max}).$ 

Estimation of the decay rate

$$\alpha > 0, P \succ 0$$

$$(K^{T}B^{T} + \Lambda^{T}(\tau)A^{T}) P\Lambda(\tau) + \Lambda^{T}(\tau) P(A\Lambda(\tau) + BK) \prec -2\alpha\Lambda^{T}(\tau) P\Lambda(\tau),$$

 $\forall \tau \in [0, h_{max}).$ 

Problem : infinite number of LMI conditions !

Estimation of the decay rate

$$\alpha > 0, P \succ 0$$

$$\begin{pmatrix} \mathsf{K}^{\mathsf{T}} \mathsf{B}^{\mathsf{T}} + \mathsf{\Lambda}^{\mathsf{T}}(\tau) \mathsf{A}^{\mathsf{T}} \end{pmatrix} \mathsf{P} \mathsf{\Lambda}(\tau) + \mathsf{\Lambda}^{\mathsf{T}}(\tau) \mathsf{P}(\mathsf{A} \mathsf{\Lambda}(\tau) + \mathsf{B} \mathsf{K}) \\ \prec -2\alpha \mathsf{\Lambda}^{\mathsf{T}}(\tau) \mathsf{P} \mathsf{\Lambda}(\tau) \,,$$

 $\forall \tau \in [0, h_{max}).$ 

- Problem : infinite number of LMI conditions !
- Solution : convex embedding with finite number of generators

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#### Exponential uncertainty

Integration operator

$$\Lambda(\rho) = e^{\rho A} + \int_{0}^{\rho} e^{sA} ds BK$$
$$= I + \int_{0}^{\rho} e^{sA} ds (A + BK)$$

Exponential uncertainty :

$$\Gamma(
ho) = \int_0^{
ho} e^{sA} ds, \ 
ho \in [0, h_{max}).$$

Use a classical representation from the robust control point of view !

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# Exponential uncertainty - Parametric uncertainty representation

$$\Gamma(
ho) = \int_{0}^{
ho(k)} e^{As} ds$$
 $0 < 
ho < h_{max}$ 



Treat the term like a parametric uncertainty

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#### Exponential uncertainty - Convex Polytope



(Hetel, Trans. Autom. Contr. 2006) (Cloosterman, Trans. Autom. Contr. 2009), (Olaru, IFAC World Congress 2007),

#### Jordan normal form

$$\lambda_{1} = -1.5, \lambda_{2} = 0.2$$
  
For  $A = \begin{pmatrix} \lambda_{1} & 0 \\ 0 & \lambda_{2} \end{pmatrix}$   
$$\Gamma(\rho) = \begin{pmatrix} \alpha_{1}(\rho) & 0 \\ 0 & \alpha_{2}(\rho) \end{pmatrix}$$
  
with  
$$\alpha_{i}(\rho) = \int_{0}^{\rho} e^{\lambda_{i}s} ds$$
  
vertex = max or min  $\alpha_{i}(\rho)$   
$$\Gamma(\rho) = \sum_{j=0}^{2^{n}} \mu_{j}A_{j}$$

## Jordan normal form + gridding

For 
$$A = \begin{pmatrix} \lambda_1 & 0 \\ 0 & \lambda_2 \end{pmatrix}$$
  
 $\Gamma(\rho) = \begin{pmatrix} \alpha_1(\rho) & 0 \\ 0 & \alpha_2(\rho) \end{pmatrix}$ 

with

$$lpha_i(
ho) = \int_0^
ho e^{\lambda_i s} ds$$

vertex = max or min  $\alpha_i(\rho)$ 

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$$\mathcal{D}_{15}^{15}$$

 $\lambda_1 = -1.5, \lambda_2 = 0.2$ 

$$\Gamma(\rho) = \sum_{j=0}^{5 \times 2^n} \mu_j A_j$$

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#### Tractable LMI conditions



Stability conditions :  $P = P^T \succ 0$ 

$$\begin{pmatrix} A^T P + PA + G_1 + G_1^T + \alpha P & PBK - G_1A_j + G_2^T \\ K^T B^T P - A_j^T G_1^T + G_2 & -G_2A_j - A_j^T G_2^T \end{pmatrix} \prec \mathbf{0},$$
  
$$\forall l = 1, \dots, N.$$

#### Numerical example

Consider a continuous-time system described by :

$$A = \begin{pmatrix} 1 & 15 \\ -15 & 1 \end{pmatrix}$$
 and  $B = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$ .

 $\blacktriangleright \lambda(A) = 1 \pm 15i$ 

▶ *K* - obtained by pole placement :

$$\lambda(A+BK)=-1\pm i$$

#### Stability analyzis comparison :

- ▶ (Mirkin, TAC 2007) : *h* ∈ [0, 0.014]
- ▶ (Naghshtabrizi, Hespanha, Teel, SCL 2008) : $h \in [0, 0.033]$
- (Fujioka, Automatica 2009) :  $h \in [0, 0.07]$
- continuous-time convex embedding :  $h \in [0, 0.09]$

#### Numerical example



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#### Conclusion and Perspective

- Robustess to sampling jitter
- Provide robust methods for stability
- Show how to reduce the conservatism of stability analyzis by taking into account the sawtooth form of the delay

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Perspective : control design